

Filed by Express Mail  
 (Receipt No. 10014204209)  
 on March 11, 1982  
 pursuant to 37 C.F.R. 1.10.  
 by [Signature]

## SPECIFICATION

### TITLE OF THE INVENTION

5 A MOBILE COMMUNICATION SYSTEM, AND A RADIO BASE STATION,  
 A RADIO APPARATUS AND A MOBILE TERMINAL

### BACKGROUND OF THE INVENTION

#### 10 (1) Field of the Invention

The present invention relates to a mobile communication system, and a radio station and a mobile terminal. Particularly, the present invention relates to a technique suitable for use in mobile communications in which communication is carried out in code division multiple access (CDMA).

#### 15 (2) Description of Related Art

CDMA is widely used as a mobile communication system. Frequency division multiple access (FDMA) or time division multiplex access (TDMA) is basically a system operated under conditions without interference between subscribers since it assigns resources (frequencies, time or the like) orthogonal to one another to subscribers (mobile terminals such as portable telephones or the like). On the other hand, CDMA is operated under conditions that subscribers are interfered with each other, thus being a system the

2011-03-15 01:10:02

channel capacity of which is expected to be improved by the statistical multiplexing effect.

Since the CDMA communication system is operated under conditions that signals of subscribers are interfered with each other as above, transmission power control (TPC) or forward error correction (FEC) or the like is used to control a transmitting signal power of each subscriber to be the minimum requirement, minimize the interference between them, thereby to maximize the channel capacity.

It is said that the channel capacity of a CDMA communication system is generally calculated by the following equation (1) (reference 1; A. Viterbi, CDMA Principles of Spread Spectrum Communication Addison-Wesley, (1995)):

$$N \doteq \frac{pg}{Eb/No_{th} \times (fs + fo) \times d} \times Dt \times \frac{Gs}{Ns} \times Lf \quad \dots (1)$$

In the above equation (1), N is a channel capacity (ch/sector/RF), pg is a processing gain, Eb/No<sub>th</sub> is a required Eb/No, fs is a self-cell interference ratio, fo is an other-cell interference ratio, d is a voice activity factor, Dt is degradation due to a transmitting power control error, Gs is a sectorization effect, Ns is the number of sectors /cells, and Lf is a loading factor. Incidentally, the

10043815.01.1102

channel capacity is considered using the following equation (2), omitting parameters not relating to the discussion:

$$N \doteq K \times \frac{Pg}{Eb/No_{th} \times (fs + fo)} \quad \dots (2)$$

In the equation (2), K represents a factor of an effect of other parameter.

The equation (2) signifies that when a signal-to-(interference + noise)power ratio (required Eb/No) required to satisfy desired communication quality (hereinafter also referred as transmission quality) increases, a transmission power of each station increases, thus interference is increases, which leads to a decrease in channel capacity N.

CDMA is a system that optimizes the communication quality and the channel capacity by TPC and FEC. However, CDMA has a disadvantage that the required Eb/No is degraded at a specific terminal moving speed because of a difference in response speed between TPC and FEC (reference 2; R. Padovani, Reverse Link Performance of IS-95 Based Cellular Systems IEEE Personal Communications, Third Quarter, (1994)).

This phenomenon occurs as follows:

In the mobile communication, there occurs

10043815.011102

fading (Reyleigh fading) that propagation loss fluctuates due to mainly interference between reflections from objects around the terminal, which fluctuates at a speed according to a moving speed of the terminal. The fluctuating speed of Reyleigh fading is characterized by a Doppler frequency  $f_D$  due to movement of the terminal as shown by the following equation (3) (reference 3; Okumura, Shinshi, "Fundamentals of Mobile Communications", The Institute of Electronics Information and Communication Engineers, (1986)):

$$f_D \doteq f_R \times V_M / C \dots (3)$$

where  $f_R$  is a radio frequency,  $V_M$  is a terminal moving speed and  $C$  is the speed of light.

It is seen that higher the radio frequency  $f_R$  and faster the terminal moving speed  $V_M$ , faster is the fluctuating speed of Reyleigh fading. Since the response speed is generally slow in TPC, it is possible to follow the propagation loss fluctuation due to fading to control the transmitting power when the Reyleigh fading fluctuation is slow (that is, when the Doppler frequency  $f_D$  is low), thus degradation of the communication quality can be prevented. However, when the Doppler frequency (that is, the fading frequency)  $f_D$  is high, the communication quality is degraded since TPC cannot follow the fading fluctuation.

10043815.011102

On the other hand, it is possible to disperse an effect of burst signal power reduction due to fading by combining FEC with interleaving (IL). However, when the fading speed is slow, there occurs signal power reduction for such a long time that the signal power reduction cannot be corrected since the IL cycle is finite, which leads to degradation of the communication quality.

For this, there occurs a phenomenon that noticeable reduction of the communication quality at the intermediate frequency  $f_D$  between the Doppler frequency  $f_D$  at which an effect of improvement of the communication quantity by TPC is obtained and the Doppler frequency  $f_D$  at which an effect of improvement by FEC is obtained (refer to Fig. 13 and the reference 2). Fig. 13 shows  $E_b/N_0$  required to satisfy the frame error rate (FER) = 1% in a reverse link (mobile terminal to base station, 9.6 kbps) in an IS-95 system. When the radio frequency = 850 MHz,  $f_D = 40$  Hz corresponds to a mobile terminal moving speed = 50.8 km/h, for example.

Since the terminal moving speed cannot be defined uniquely, the system has to be designed at the Doppler frequency at which the quality degradation is maximum [namely, the required  $E_b/N_0$  is maximum (in the example in Fig. 13,  $f_D = 47$  Hz and the required  $E_b/N_0 = 6.1$  dB)]. However, a large required  $E_b/N_0$  degrades

the channel capacity  $N$  as expressed by the above equation (2). On the other hand, when the channel capacity  $N$  is secured, the required  $E_b/N_0$  cannot be satisfied, which leads to degradation of the communication quality.

In a system such as TDMA or FDMA in which the transmission quality monotonically degrades when the fading pitch ( $f_D$ ) increases, an attempt is made to prevent large degradation of the transmission quality of a specific terminal to equalize the communication quality in the entire system, thereby improving the overall performance, as disclosed in, for example, Japanese Patent Laid-Open Publication No. 5-259969.

In concrete, the technique disclosed in Japanese Patent Laid-Open Publication No. 5-259969 (hereinafter referred as known technique 1) accomplishes the above effect by a simple control based on only magnitude of the terminal moving speed on the assumption of monotonousness of the Doppler frequency  $f_D$  and the transmission quality degradation that degradation of the transmission quality decreases when the fading pitch becomes smaller. However, such a simple control can provide only a small effect when degradation of the transmission quality to the Doppler frequency is non-monotonous as in CDMA (when there is employed a communication system having a characteristic that a required signal-to-noise power

20110518001

ratio of a received signal in a mobile terminal changes from a tendency to increase to a tendency to decrease according to the moving speed of the mobile terminal), which sometimes causes degradation of the communication quality in some cases.

Namely, the known technique 1 tries to assure the communication quality of a control channel by assigning a low frequency band to the control channel such that  $fD$  ( $\propto$  (moving speed)  $\times$  (radio frequency)) of the control channel required a high communication quality becomes as low as possible. However, when the Doppler frequency  $fD$  is lowered in a system such as CDMA in which degradation of the transmission quality to the Doppler frequency  $fD$  is non-monotonous in the similar manner, there is a case where the communication quality degrades as with the case where  $fD > 42$  Hz in Fig. 13, for example.

The known technique 1 accommodates traffic of a high-speed mobile terminal in a low radio frequency band and a low-speed moving terminal in a high radio frequency band to equalize the Doppler frequency  $fD$  ( $\propto$  (moving speed)  $\times$  (radio frequency)) as much as possible, thereby homogenizing the transmission quality to improve the overall performance.

In this method, when a radio frequency band in which the Doppler frequency brings about the worst value of the transmission quality ( $fD = 42$  Hz in the

10043815.011102

example in Fig. 13) is assigned, the transmission quality of the both terminals becomes the worst, thus the overall performance becomes the worst. The known technique 1 cannot prevent this.

5           The above problem is caused by that the known technique 1 performs a simple control based on only magnitude of the terminal moving speed on the assumption of monotonousness of the Doppler frequency  $f_D$  and the transmission quality degradation.

10           In consideration of assignment of frequencies in a CDMA communication system, there is a technique disclosed in, for example, Japanese Patent Laid-Open Publication No. 10-23502 (hereinafter referred as known technique 2) as a known technique. An objective  
15 of the known technique 2 is to equalize the number of terminals to be accommodated in each radio frequency band (strictly, the number of radio channels to be used) in order to secure a channel for soft hand-off in the CDMA communication system.

20           However, the known technique 2 does not consider non-monotonousness or the like of the transmission quality to the terminal moving speed and the Doppler frequency  $f_D$ , but assigns a radio frequency band without considering the terminal moving speed.  
25 For this, when the Doppler frequency  $f_D$  brings about the worst value of the transmission quality at the assigned frequency ( $f_D = 42$  Hz in the example in Fig.

10043845.011102



13), large degradation occurs.

#### SUMMARY OF THE INVENTION

5           In the light of the above problems, an object  
of the present invention is to improve communication  
quality and increase a channel capacity in a mobile  
communication system in which a relationship between  
a terminal moving speed (Doppler frequency) and  
10 transmission quality degradation is non-monotonous.

          The present invention therefore provides a  
mobile communication system comprising a detecting  
unit to detect information concerning a moving speed  
of the mobile terminal (hereinafter referred as a  
15 terminal moving speed) on the basis of a received signal  
from the mobile terminal, and a selection controlling  
unit to select the use frequency in a higher radio  
frequency band when the information detected by the  
detecting unit is higher, selecting the use frequency  
20 in a lower radio frequency band when the information  
is lower, and assigning the selected use frequency to  
the communication between the mobile terminal and the  
radio base station.

          A radio base station according to this  
25 invention for realizing the above mobile communication  
system comprises:

(1) a radio communicating unit being able to

20130305 0440

communicate with the mobile terminal using any one of  
M (M being an integer not less than two) radio frequency  
bands;

(2) a speed information detecting unit to  
5 detect information concerning a moving speed of the  
mobile terminal (herein after referred as speed  
information) on the basis of a received signal from  
the mobile terminal received by the radio communicating  
unit; and

10 (3) a use frequency selection controlling unit  
to select the use frequency in a higher radio frequency  
band when the speed information detected by the speed  
information detecting unit is higher, select the use  
frequency in a lower radio frequency band when the speed  
15 information is lower, and assign the selected use  
frequency to the communication with the mobile  
terminal.

By providing the above units to the radio base  
station, it is possible to switch the use radio  
20 frequency band at specific terminal moving speed  
information to improve the worst value of the required  
signal-to-noise power ratio.

A radio apparatus according to this invention  
being able to use both a frequency belonging to a first  
25 frequency band and a frequency belonging to a second  
frequency band higher than the first frequency band  
for communication on forward and reverse links with

10043815 011102

a mobile terminal, the radio apparatus comprising:

(1) a transmitting unit to convert a signal obtained by error-correction-encoding and interleave transmitting data into a radio signal, and transmit  
5 the radio signal for communication on the forward link to the mobile terminal;

(2) a transmitting power controlling unit to control a transmitting power of the radio signal for communication on the forward link on the basis of a  
10 received signal from the mobile terminal; and

(3) a selection controlling unit to use a frequency belonging to the second frequency band for communication with the mobile terminal when determining that a fading cycle of the received signal  
15 from the mobile terminal or a moving speed of the mobile terminal is fast, use a frequency belonging to the first frequency band for communication with the mobile terminal when determining that the fading cycle or the moving speed of the mobile terminal is slow.

20 A radio apparatus according to this invention being able to use both a frequency belonging to a first frequency band and a frequency belonging to a second frequency band higher than the first frequency band for communication on forward and reverse links with  
25 a mobile terminal, the radio apparatus comprising:

(1) a transmitting unit to convert a signal obtained by encoding and interleaving transmitting

10043815.011102

data into a radio signal, and transmit the radio signal for communication on the forward link to the mobile terminal;

(2) a transmitting power controlling unit to  
5 control a transmitting power of the radio signal for communication on the forward link on the basis of a received signal from the mobile terminal; and

(3) a selection controlling unit to use a frequency belonging to the second frequency band in  
10 communication with the mobile terminal when determining on the basis of the received signal from the mobile terminal that a fading cycle in a received signal on the forward link received by the mobile terminal or a moving speed of the mobile terminal is  
15 fast, use a frequency belonging to the first frequency band in communication with the mobile terminal when determining that the fading cycle or the moving speed of the mobile terminal is slow.

A reference to determine whether a speed of  
20 the mobile terminal is fast or slow is determined on the basis of a required signal-to-noise power ratio of each of the above frequency bands. For example, when a communication system having a characteristic that the required signal-to-noise power ratio of a  
25 received signal in the terminal changes from a tendency to increase to a tendency to decrease according to a moving speed of the mobile terminal is employed, a

10043315.01102

terminal moving speed satisfying the required signal-to-noise power ratio of each frequency band may be the reference.

Since the above Doppler frequency  $f_D$  is  
5 expressed by the above equation (3), the Doppler frequency  $f_D$  changes according to the use radio frequency even at the same terminal moving speed, thus the required signal-to-noise power ratio changes. Therefore, a terminal moving speed at which the  
10 required signal-to-noise power ratio is maximum differs from use radio frequency to use radio frequency. However, the use radio frequency band is switched at a specific terminal moving speed as a boundary, the worst value of the required signal-to-noise power ratio  
15 can be improved.

Since the channel capacity of mobile communication depends on not only a required signal-to-noise power ratio but also interference power information such as a self-cell interference  
20 ratio ( $f_s$ ), an other-cell interference ratio ( $f_o$ ) or the like as shown by the equation (2), it is desirable that a selection (switching) reference (threshold value information) for the use radio frequency band is determined, adding interference power information  
25 to the communication with the mobile terminal.

In such case, if the above interference power information is determined on the basis of a signal

transmission characteristic of each of the above radio frequency bands, it is possible to determine the above reference (threshold value information), considering fluctuation in radio wave propagation loss and the like,  
5 for example.

A mobile terminal according to this invention for realizing the above mobile communication system comprising:

(1) a radio communicating unit being able to  
10 communicate with the radio base station using any one of M (M being an integer not less than two) radio frequency bands;

(2) a selected frequency notification signal receiving unit to receive, from the radio communicating  
15 unit, a selected frequency notification signal for notifying of a use frequency selected among higher radio frequency bands in the radio base station when speed information of its own is faster or selected among lower radio frequency bands when the speed information  
20 of its own is slower; and

(3) a use frequency selection controlling unit to select a radio frequency to be used in the radio communicating unit among the radio frequency bands according to the selected frequency notification  
25 signal received by the selected frequency notification signal receiving unit.

With the above structure, the mobile terminal

according to this invention selects a radio frequency to be used in the communication with the radio base station in each of the above radio frequency bands according to the selected frequency notification  
5 signal for notifying of the radio frequency selected as above in the radio base station, thereby improving the worst value of the required signal-to-noise power ratio and the channel capacity in the whole system.

The present invention provides the following  
10 advantages and effects:

(1) As a radio frequency to be used (assigned) in the communication between the radio base station (radio apparatus) and the mobile terminal, the use frequency is selected in a higher radio frequency band  
15 when the speed information concerning a moving speed of the mobile terminal is higher, while the use frequency is selected in a lower radio frequency band when the speed information is lower. It is thus possible to improve the worst value of the required  
20 signal-to-noise power ratio in a mobile communication system in which a relationship between the Doppler frequency and transmission quality degradation is non-monotonous (employing a communication system having a characteristic that the required  
25 signal-to-noise power ratio of a received signal in the mobile terminal changes from a tendency to increase to a tendency to decrease according to a moving speed

of the mobile terminal), improve the communication quality, and increase the channel capacity.

(2) When the threshold value information for the speed information that is a reference (boundary) between the above higher speed and lower speed is determined adding interference power information to the communication with the mobile terminal, threshold value information meeting actual communication environments can be obtained, which leads to further improvement of the communication quality and channel capacity.

(3) If the above interference power information is determined on the basis of a signal transmission characteristic of each of plural radio frequency bands, it is possible to determine the above threshold value in consideration of fluctuation in radio wave propagation loss and the like. Accordingly, improvement of the worst value of the required signal-to-noise power ratio may bring more effects of improvement of the channel capacity.

(4) The above interference power information may be calculated on the basis of information on the number of mobile terminals presently in communication and an actual measured value of the received signal-to-noise power. In such case, even if the interference power information fluctuates due to fluctuation in system operational environments,

201403215.011102



fluctuation in base station installation conditions  
or fluctuation in traffic with time, optimal threshold  
value information can be re-calculated following the  
fluctuation. It is therefore possible to  
5 appropriately switch the frequency in consideration  
of fluctuation in actual system operational  
environments, in base station installation conditions,  
intraffictwithtimeandthelike. Accordingly, a large  
improvement of the communication quality and a large  
10 increase in channel capacity are expected.

(5) The above threshold value information may  
be prepared ( $2 \times M - 1$ ) pieces, and the use radio frequency  
may be selected on the basis of information on which  
range of the threshold value information speed  
15 information on the mobile terminal falls in, and  
priority information defining which radio frequency  
band should be used in each of the plural terminal speed  
ranges defined by each piece of the threshold value  
information. In such case, it is possible to alleviate  
20 concentration of calls in a specific radio frequency  
band without causing degradation of the worst value  
of the required signal-to-noise power ratio.

#### BRIEF DESCRIPTION OF THE DRAWINGS

25

Fig. 1 is a block diagram showing a structure  
of a CDMA communication system (mobile communication

100433815 044102

system) according to an embodiment of this invention;

Fig. 2 is a block diagram showing in detail structures of essential parts of a CDMA transmitting unit in a radio base station apparatus shown in Fig.

5 1;

Fig. 3 is a block diagram showing in detail structures of essential parts of a CDMA receiving unit in the radio base station apparatus shown in Fig. 1;

Fig. 4 is a block diagram showing in detail structures of essential parts of a CDMA transmitting/receiving unit in a mobile terminal apparatus shown in Fig. 1;

Fig. 5 is a graph showing an example of required Eb/No characteristics to a mobile terminal moving speed according to this embodiment;

Fig. 6 is a sequence diagram for illustrating an operation of the CDMA communication system shown in Fig. 1;

Fig. 7 is a graph showing an example of characteristics of "a product of Eb/No<sub>th</sub> and interference power information (fs+fo)" to a mobile terminal moving speed according to a first modification of this embodiment;

Fig. 8 is a block diagram showing a structure of a radio base station apparatus according to a second modification of this embodiment;

Fig. 9 is a block diagram showing a structure

2011052640T 10043815 011102

of a radio base station apparatus according to a third modification of this embodiment;

Fig. 10 is a diagram for illustrating a manner of setting a plurality of switching speed threshold values (when  $M=2$ ) in the radio base station apparatus shown in Fig. 9;

Fig. 11 is a diagram for illustrating a manner of setting a plurality of switching speed threshold values (when  $M=3$ ) in the radio base station apparatus shown in Fig. 9;

Fig. 12 is a sequence diagram for illustrating an operation of a CDMA communication system according to the third modification; and

Fig. 13 is a graph showing an example of a required  $E_b/N_0$  characteristic to a fading frequency in a CDMA communication system.

#### DESCRIPTION OF THE PREFERRED EMBODIMENTS

Hereinafter, description will be made of embodiments of this invention with reference to the drawings.

##### (A) Description of an Embodiment

Fig. 1 is a block diagram showing a structure of a CDMA communication system (mobile communication system) according to an embodiment of this invention. The CDMA communication system shown in Fig. 1 comprises

at least one radio base station apparatus (radio apparatus; hereinafter referred simply as a base station) 1, and at least one mobile terminal apparatus (hereinafter referred simply as a terminal) 2. The  
5 base station 1 can make a two-way radio communication in CDMA system with a terminal (in-zone terminal) present in its own communicable area (cell).

#### (A1) Base Station Structure

In concrete, the base station 1 according to  
10 this embodiment comprises, when attention is paid to its essential parts, a multiplexing unit 11, a change-over switch 12, a CDMA transmitting unit 13a for fR1 band (first frequency band) and a CDMA transmitting unit 13b for fR2 band (second frequency  
15 band) as a transmitting system, a CDMA receiving unit 16a for fR1 band, a CDMA receiving unit 16b for fR2 band, a change-over switch 17 and a demultiplexing unit 18 as a receiving system, as shown in, for example, Fig. 1. The base station 1 further comprises a  
20 change-over switch 19, a Doppler frequency (fD) measuring unit 20A, a terminal moving speed estimating unit 20B, a switching speed threshold value holding unit 20C and a frequency band switching controlling unit 20D as a control system 20.

25 Although not shown in Fig. 1, the base station 1 has a TPC function for performing TPC (transmitting power control) on the forward link (radio signals for

forward communication) on the basis of a received signal from the terminal 2. In Fig. 1, reference characters 14a and 14b denote hybrids (H), 15a and 15b denote transmitting/receiving antennae for fR1 and fR2 band, respectively, which are shared by the above transmitting system and the receiving system.

Namely, a transmitting signal from the CDMA transmitting unit for fR1 band (fR2 band) 13a (13b) is transmitted toward the terminal 2 from the transmitting/receiving antenna 15a (15b) through the hybrid 14a (14b). On the other hand, a radio signal in the fR1 band (the fR2 band) from the terminal 2 is received by the transmitting/receiving antenna 15a (15b), and outputted to the CDMA receiving unit for fR1 band (fR2 band) 16a (16b) by the hybrid 14a (14b).

Namely, the CDMA transmitting units 13a and 13b, the hybrids 14a and 14b, the CDMA receiving units 16a and 16b, and the transmitting/receiving antennae 15a and 15b function as a radio communicating unit, which can communicate with the terminal 2 using either one of M=2 types of radio frequency bands (the fR1 band, the fR2 band).

#### (A1.1) Transmitting System Structure

In the above transmitting system, the multiplexing unit (control signal adding unit) 11 multiplexes (adds) transmitting data to the terminal 2 and a control signal (a selected frequency

10043815-011102

notification signal, a switching timing instruction signal or the like) from the frequency band switching controlling unit 20D. The change-over switch 12 switches its output according to a switching  
 5 controlling signal from the frequency band switching controlling unit 20D to supply the output (a transmitting signal) from the multiplexing unit 11 to either the CDMA transmitting unit for fR1 band 13a or the CDMA transmitting unit for fR2 band 13b.

10 The CDMA transmitting unit 13a for fR1 band outputs the transmitting signal from the change-over switch 12 as a signal in a specific frequency band (the fR1 band). The CDMA transmitting unit 13b for fR2 band outputs the transmitting signal from the change-over  
 15 switch 12 as a signal in another frequency band (the fR2 band) differing from the above fR1 band. Incidentally, a relationship in height between fR1 (the first frequency band) and fR2 (the second frequency band) is  $fR2 > fR1$  in this embodiment; the fR1 band =  
 20 850-MHz band, and the fR2 band = 2-GHz band, for example.

#### • CDMA Transmitting Unit Structure

Each of the above CDMA transmitting units 13a and 13b has a common structure in view of hardware, comprising traffic channel transmission processing  
 25 units 131 equal in number to traffic channels, a multiplexing unit (MUX) 132, a scrambler (multiplier) 133, a digital filter 134, a DA converter (DAC:

10043815, 01-11-02

Digital-to-Analog Converter) 135, an orthogonal modulator 136, an RF (Radio Frequency) oscillator for forwardlink (base station 1 to terminal 2) 137, a filter 138, a high-power amplifier (HPA) 139 and the like.

- 5 An oscillated frequency of the above RF oscillator 137 is changed by each of the CDMA transmitting units 13a and 13b, whereby the CDMA transmitting unit 13a can generate a transmitting signal in the fR1 band, while the CDMA transmitting unit 13b can generate a  
10 transmitting signal in the fR2 band.

- In concrete, a transmitting signal [(transmitting data) + (control signal)] is undergone necessary FEC (error correction) coding and interleaving process in an FEC encoder 131a and an  
15 interleaver 131b, multiplied by a channelisation code such as a Walsh code or the like in a multiplier 131c to be a channelisation signal, and multiplexed on a pilot channel signal and a control channel signal in the multiplexing unit 132.

- 20 The multiplexed signal obtained as above is scrambled (spectrum-spread-modulated) with a scrambling code that is a PN (pseudo noise) code (chip rate = 1.2288 Mcps) in the scrambler 133, filtered (removed unnecessary signal components thereof) in the  
25 digital filter 134, DA-converted in the DA converter 135, undergone orthogonal-modulation such as QPSK or the like with an RF signal from the RF oscillator 137

in the orthogonal modulator 136, finally outputted to the hybrid 14a or 14b shown in Fig. 1 through the filter 138 and the HPA 139 to be transmitted toward the terminal 2 from the transmitting/receiving antenna 15a or 15b.

5           In the structure shown in Fig. 2, the transmitting signal is modulated in the RF band. However, various manners are possible. For example, the transmitting signal may be modulated in the intermediate frequency (IF) band, then  
10 frequency-converted to the RF band (up-converted).

#### (A1.2) Receiving System Structure

          In Fig. 1, the CDMA receiving unit for fR1 band 16a performs a necessary demodulating process on a received signal in the fR1 band from the terminal 2  
15 inputted through the hybrid 14a. The CDMA receiving unit for fR2 band 16b performs a necessary demodulating process on a received signal in the fR2 band from the terminal 2 inputted through the hybrid 14b.

#### • CDMA Receiving Unit Structure

20           Each of the CDMA receiving units 16a and 16b according to this embodiment comprises a low noise amplifier (LNA) 161, a filter 162, traffic channel reception processing units 163 equal in number to the traffic channels, and the like, as shown in, for example,  
25 Fig. 3. Each of the traffic channel receiving units 163 comprises an orthogonal detector 163a, an RF oscillator for reverse link (terminal 2 to base station



1) 163b, an AD converter (ADC: Analog-to-Digital Converter) 163c, a descrambler (multiplier) 163d, a multiplier 163e, an M-Array correlating detector 163f, a deinterleaver 163g, an error correction decoder 163h, and the like. Like a relationship between the CDMA transmitting units 13a and 13b, an oscillated frequency of the above RF oscillator 163b is changed by each of the CDMA receiving units 16a and 16b, whereby the CDMA receiving unit 16a can process a received signal in the fR1 band, while the CDMA receiving unit 16b can process a received signal in the fR2 band.

In concrete, the received signal from the terminal 2 is amplified and filtered in the LNA 161 and the filter 162, then inputted to each of the traffic channel reception processing units 163.

In each of the traffic channel reception processing units 163, the inputted signal (received signal) is orthogonal-detected with the RF signal from the RF oscillator 163b in the orthogonal detector 163a, AD-converted in the AD converter 163c, and undergone a descrambling (despreading) process with a PN code and a channel demodulating process with a channelisation code in the descrambler 163d and the multiplier 163e.

Thereafter, the received signal is undergone an M-Array correlating detecting process in the M-Array correlating detector 163f, undergone a deinterleaving

10043815.011102

process and an error correction decoding process in the deinterleaver 163g and the error correction decoder 163h, and outputted to the demultiplexing unit 18 through the above change-over switch 17.

5 • M-Array Correlating Detection

Here, description will be made of the above M-Array correlating detecting process.

In a land mobile communication system such as a cellular system, the amplitude or phase of a transmitting signal rapidly fluctuates due to an effect of fading, so that it is impossible to perform coherent detection to regenerate a reference phase from the received signal and demodulate it. For this reason, a CDMA communication system in conformity with IS-95  
10 transmits a pilot channel on the forward link, the receiving side (terminal 2) performs coherent detection with the pilot channel as the reference phase, while adopting the M-Array correlating detection system on the reverse link.  
15

20 Incidentally, although the IS-95 system uses the M-Array correlating detecting system with Walsh codes having a length of 64 bits, here is described a case in which Walsh codes having a length of 16 bits (16 to 4-bit M-Array) are used as shown in the following  
25 table 1 will be here described, for the sake of simplicity.

10043815 011102

[Table 1]

TABLE OF 16-BIT WALSH CODES

| CODE NUMBER<br>(BINARY NUMERAL) | WALSH CODE          |
|---------------------------------|---------------------|
| W0 (0000)                       | 0000 0000 0000 0000 |
| W1 (0001)                       | 0101 0101 0101 0101 |
| W2 (0010)                       | 0011 0011 0011 0011 |
| W3 (0011)                       | 0110 0110 0110 0110 |
| W4 (0100)                       | 0000 1111 0000 1111 |
| W5 (0101)                       | 0101 1010 0101 1010 |
| W6 (0110)                       | 0011 1100 0011 1100 |
| W7 (0111)                       | 0110 1001 0110 1001 |
| W8 (1000)                       | 0000 0000 1111 1111 |
| W9 (1001)                       | 0101 0101 1010 1010 |
| W10 (1010)                      | 0011 0011 1100 1100 |
| W11 (1011)                      | 0110 0110 1001 1001 |
| W12 (1100)                      | 0000 1111 1111 0000 |
| W13 (1101)                      | 0101 1010 1010 0101 |
| W14 (1110)                      | 0011 1100 1100 0011 |
| W15 (1111)                      | 0110 1001 1001 0110 |

It is clearly seen from Table 1 that eight bits  
 5 in each code differ from the others in other code (code  
 distance = 8). The transmitting side (terminal 2)  
 collects four bits of data to be transmitted  
 (interleaved output), and selects a Walsh code  
 corresponding to the binary digit value. For instance,  
 10 when the interleaved output is "1011", a code number  
 = W11 (0110 0110 1001 1001) is selected (16 to 4-bit  
 M-Array encoding), and transmitted.

On the other hand, the receiving side (base  
 station 1) collects every 16 bits of received data,  
 15 and compares them with each of all the Walsh codes  
 (W0-W15) to determine whether they coincide with each  
 other. The receiving side determines that a Walsh code

10043815 011102  
 2011052001

having the largest number of coinciding bits is a code actually transmitted, and outputs its number (binary number four bits) as a result of correlation detection. For instance, assuming that an error of three bits occurs while a Walsh code (0110 0110 1001 1001) transmitted from the terminal 2, and the base station 1 receives a code (0100 0111 1001 1000).

When this code is compared with each of the code numbers W0-W15, the results are as shown in the following table 2.

[Table 2]

RESULTS OF COMPARISON

| CODE NUMBER<br>(BINARY NUMERAL) | WALSH CODE          | THE NUMBER OF<br>COINCIDING<br>BITS |
|---------------------------------|---------------------|-------------------------------------|
| W0 (0000)                       | 0000 0000 0000 0000 | 9                                   |
| W1 (0001)                       | 0101 0101 0101 0101 | 9                                   |
| W2 (0010)                       | 0011 0011 0011 0011 | 7                                   |
| W3 (0011)                       | 0110 0110 0110 0110 | 7                                   |
| W4 (0100)                       | 0000 1111 0000 1111 | 9                                   |
| W5 (0101)                       | 0101 1010 0101 1010 | 9                                   |
| W6 (0110)                       | 0011 1100 0011 1100 | 7                                   |
| W7 (0111)                       | 0110 1001 0110 1001 | 7                                   |
| W8 (1000)                       | 0000 0000 1111 1111 | 7                                   |
| W9 (1001)                       | 0101 0101 1010 1010 | 11                                  |
| W10 (1010)                      | 0011 0011 1100 1100 | 9                                   |
| W11 (1011)                      | 0110 0110 1001 1001 | 13                                  |
| W12 (1100)                      | 0000 1111 1111 0000 | 11                                  |
| W13 (1101)                      | 0101 1010 1010 0101 | 7                                   |
| W14 (1110)                      | 0011 1100 1100 0011 | 5                                   |
| W15 (1111)                      | 0110 1001 1001 0110 | 9                                   |

As seen from Table 2, the code number W11 has the largest number (13) of coinciding bits, so that a code number W11 is selected, and "1011" is outputted

as a result of the correlation detection. In this case, since a code distance of the Walsh codes is eight, it is possible to accurately detect even if an error of up to three bits occurs in 16 bits.

5           The actual IS-95 system uses Walsh codes having a length of 64 bits, thus an input of the correlating detector 163f is 64 bits, while an output of the same is six bits ( $26=64$ ) (namely, 64 to 6-bit M-Array), as shown in Fig. 3. In this case, since the code distance  
10 of a 64-bit Walsh code is 32, it is possible to decode even if an error of up to 15 bits occurs in 64 bits.

In the practical system, a process for extracting soft decision information for error correction decoding is often added. In CDMA 2000 or  
15 the like, coherent detection using a pilot channel is performed even on the reverse link.

In the change-over switch 17 in Fig. 1, an input thereof is switched according to a switching controlling signal from the frequency band switching  
20 controlling unit 20D, whereby the change-over switch 17 outputs either one of outputs from the above CDMA receiving units 16a and 16b (outputs from the error correction decoders 163h) to the demultiplexing unit 18. The demultiplexing unit (confirmation signal  
25 extracting unit) 18 demultiplexes the output (received signal) from the change-over switch 17 into received data and a control signal (reception confirmation

signal in response to a selected frequency notification signal, a confirmation signal in response to a switching timing instruction signal, or the like to be described later) from the terminal 2. Incidentally, 5 the control signal from the terminal 2 is supplied to the frequency band switching controlling unit 20D (switching timing instruction signal generating unit to be described later).

#### (A1.3) Control System Structure

10 In the change-over switch 19, an input thereof is switched according to a switching control signal from the frequency band switching controlling unit 20D, whereby the change-over switch 19 supplies either one of outputs of information useful to measure (detect) 15 the Doppler frequency  $f_D$  obtained in the CDMA receiving units 16a or 16b to the Doppler frequency measuring unit 20A. The Doppler frequency measuring unit (speed information detecting unit; detecting unit) 20A measures (detects) the Doppler frequency  $f_D$  20 (information on a moving speed of the terminal 2) of a received signal from the received signal inputted from the change-over switch 19.

The terminal moving speed estimating unit 20B estimates a moving speed of the terminal 2 that has 25 transmitted the received signal on the basis of the Doppler frequency  $f_D$  obtained by the above Doppler frequency measuring unit 20A. The Doppler frequency

measuring unit 20A and the terminal moving speed estimating unit 20B can be configured, by applying a technique described in, for example, Japanese Patent Laid-Open Publication No. 6-242225 or the like.

5           In this case, an output of the AD converter 163c (an output of the orthogonal detector 163a) may be used as useful information to measure (detect) the above Doppler frequency  $f_D$ . Alternatively, when AGC (Automatic Gain Control) is performed in the CDMA  
10 receiving units 16a and 16b, gain information thereof may be used, or an output of the M-Array correlating detector 163f (a result of correlating detection) may be used, or the both may be used for improvement of accuracy of measurement of the Doppler frequency  $f_D$ .

15           The switching speed threshold value holding unit 20C holds a threshold value (hereinafter referred as a switching speed threshold value) for a terminal moving speed that becomes a reference when the change-over switches 12, 17 and 19 are switched. The  
20 switching speed threshold value holding unit 20C is configured with a RAM or the like, for example. The switching speed threshold value may be suitably set by a host system managing operation of the entire base station 1.

25           The frequency band switching controlling unit (use frequency selection controlling unit; selection controlling unit) 20D compares the terminal moving

speed obtained by the terminal moving speed estimating unit 20B with the above switching speed threshold value, controls the change-over switches 12, 17 and 19 according to a result of the comparison, and selects  
 5 (switches) a radio frequency band (the fR1 band or the fR2 band) to be used in the communication with the terminal 2.

Here is described a manner of determining the above switching speed threshold value. Fig. 5 is a  
 10 graph showing an example of required  $E_b/N_0$  (signal-to-(interference + noise) power ratios required to satisfy desired communication quality) to the terminal moving speed when two kinds of radio frequency bands (the fR1 band and the fR2 band) are  
 15 850-MHz band and 2-GHz band, respectively. The graph is obtained by calculating the Doppler frequency  $f_D$  to the terminal moving speed using the above equation (3) for each of the radio frequency bands, applying a result of the calculation to Fig. 13, and determining  
 20 a required  $E_b/N_0$ . In Fig. 5, a characteristic 31 represents a required  $E_b/N_0$  characteristic at a radio frequency in the 850-MHz band, while a characteristic 32 represents a required  $E_b/N_0$  characteristic at a radio frequency in the 2-GHz band.

25 As seen from Fig. 5, when a radio frequency in the 850-MHz band is used, the required  $E_b/N_0$  is maximum (the worst value) (the required  $E_b/N_0$  changing



from a tendency to increase to a tendency to decrease)  
at a point (at the terminal moving speed of about 60  
km/h) denoted by a reference numeral 31a. When a radio  
frequency in the 2-GHz band is used, the required  $E_b/N_0$   
5 is of the worst value at a point (at the terminal moving  
speed of about 25 km/h) denoted by a reference numeral  
32a. Incidentally, a non-monotonous characteristic  
as this is peculiar when a CDMA system using both TPC  
and FEC is employed, as described above.

10 Accordingly, if the use radio frequency band  
is switched at a point corresponding to the maximum  
value (about 5.8 dB) of the required  $E_b/N_0$  that the  
above both characteristics 31 and 32 (that is, an  
intersection of the characteristics 31 and 32) are  
15 satisfied as a boundary (namely, a switching speed  
threshold value), it is possible to improve the worst  
value of the required  $E_b/N_0$ .

Namely, when the terminal moving speed is below  
42.5 km/h, a radio frequency belonging to the 850-MHz  
20 band is selected as a radio frequency to be used to  
communicate with the base station 1. When the terminal  
moving speed exceeds it, a radio frequency belonging  
to the 2GHz-band is selected. In other words, the  
higher the terminal moving speed, the use frequency  
25 is selected from a higher frequency band (2-GHz), and  
assigned for the communication with the terminal 2.  
Conversely, the lower the terminal moving speed, the

use frequency is selected from a lower frequency band (850-MHz).

As this, the switching speed threshold value is determined on the basis of the required  $E_b/N_0$  characteristics 31 and 32 relating to the radio frequency band and the terminal moving speed, the use radio frequency band is switched according to a result of comparison between the switching speed threshold value and the terminal moving speed. Whereby, the worst value of the required  $E_b/N_0$  can be improved to about 5.8 dB in this embodiment.

Assuming that the worst values of the required  $E_b/N_0$ s in the 850-MHz band and 2-GHz band are about 6.1 dB, a quantity of the improvement becomes about  $6.1 - 5.8 = 0.3$  dB (7.2 % when it is converted into a quantity of improvement of the channel capacity). Meanwhile, since there is a possibility to obtain a larger effect of the improvement depending on the terminal moving speed, it is expected that an actual quantity of improvement of the channel capacity is becomes larger. A practical switching speed threshold value or an effect of the improvement changes according to a used radio frequency, TPC system or FEC system. It is thus possible to readily determine them by making a graph corresponding to Fig. 5 in simulation or the like when the characteristics are determined.

In this case, the frequency band switching

controlling unit 20D such controls the change-over switches 12, 17 and 19 that the use radio frequency band is the 850-MHz band when the terminal moving speed is not higher than 42.5 km/h, or the 2-GHz band when  
5 the terminal moving speed exceeds it. When the base station 1 switches the use radio frequency band, the communication cannot be continued if the terminal 2 does not switch the use radio frequency band to the same radio frequency band as the base station 1.

10 Therefore, it is necessary to synchronize the use radio frequency between the base station 1 and the terminal 2. According to this embodiment, a radio frequency band (to be switched to) and a switching timing selected on the side of the base station 1 is  
15 notified to the terminal 2 with a control signal (selected frequency notification signal, switching timing instruction signal), whereby synchronization is established.

For this purpose, the frequency band switching  
20 controlling unit 20D according to this embodiment also has a function as a notification signal generating unit for generating the above selected frequency notification signal, and a switching timing instruction signal generating unit for generating a  
25 switching timing instruction signal. The latter switching timing instruction signal is generated when a reception confirmation signal in response to the

former selected frequency notification signal is received from the terminal 2 (when the reception confirmation signal is demultiplexed from the received signal in the demultiplexing unit 18) as an opportunity, and multiplexed on transmitting data in the multiplexing unit 11 as well as the former selected frequency notification signal.

#### (A2) Moving Terminal Structure

Next description will be made of a structure of the terminal 2 according to this embodiment. The terminal 2 according to this embodiment comprises, when attention is paid to essential parts thereof, a CDMA transmitting/receiving unit 23 for fR1 band, a CDMA transmitting/receiving unit 24 for fR2 band, change-over switch 25, a control signal multiplexing/demultiplexing unit 26 and a switching controlling unit 27, as shown in Fig. 1.

The CDMA transmitting/receiving unit 23 for fR1 band makes a radio communication with the base station 1 in CDMA system using a radio signal in a radio frequency band in the fR1 band (for example, 850-MHz or the like), whereas the CDMA transmitting/receiving unit 24 for fR2 band makes a radio communication with the base station 1 in CDMA system using a radio signal in a radio frequency band in the fR2 band (for example, the 2-GHz band). Namely, these CDMA transmitting/receiving units 23 and 24 function as a

2014032815:011102

radio communicating unit which can communicate with the base station 1 using either one of two kinds ( $M=2$ ) of radio frequency bands (the  $fR1$  band and the  $fR2$  band) correspondingly to a kind of the radio frequency band used by the base station 1.

In concrete, the CDMA transmitting/receiving units 23 and 24 according to this embodiment have a common hardware structure, as shown in, for example, Fig. 4. Each of the CDMA transmitting/receiving units 23 and 24 comprises a low noise amplifier (LNA) 281, a filter 282, an orthogonal detector 283, an RF oscillator for forward link 283a, an AD converter (ADC) 284, a descrambler (multiplier) 285, a multiplier 286, an integrator (filter) 287, a deinterleaver 288, an FEC decoder 289 and the like as a receiving system 28, and an FEC encoder 291, an interleaver 292, an M-Array correlating encoder 293, a multiplier 294, a scrambler (multiplier) 295, a digital filter 296, a DA converter (DAC) 297, an orthogonal modulator 298, an RF oscillator for reverse link 298a, a filter 299, a high-power amplifier (HPA) 300 and the like as a transmitting system 29, along with a hybrid (H) 30a and a transmitting/receiving antenna 30b shared by the transmitting system 28 and the receiving system 29.

Oscillated frequencies of the RF oscillators 283a and 298a in the above receiving system 28 and transmitting system 29 are made different from each

other between the CDMA transmitting/receiving unit 23 and the CDMA transmitting/receiving unit 24, whereby the CDMA transmitting unit 23 can make a two-way radio communication with the base station 1 using a radio  
5 signal in the fR1 band, whereas the CDMA transmitting/receiving unit 24 can make a two-way radio communication with the base station 1 using a radio signal in the fR2 band.

In concrete, a received signal from the base  
10 station 1 is inputted to the receiving system 28 through the transmitting/receiving antenna 30b and the hybrid 30a, amplified and filtered by the LNA 281 and the filter 282, and undergone orthogonal detection with an RF signal (an RF signal in the fR1 band in the case of  
15 the CDMA transmitting/receiving unit 23, or an RF signal in the fR2 band in the case of the transmitting/receiving unit 24) from the RF oscillator 283a in the orthogonal detector 283. The received signal undergone the quadrature detection is converted  
20 into a digital signal in the AD converter 284, then despread and channel-decoded.

Namely, an output of the AD converter 284 is multiplied by a scrambling code (PN code) to be descrambled (despread) in the descrambler 285,  
25 multiplied by a channelisation code to be undergone the channel decoding process in the multiplier 286, and integrated in the integrator 287. The integrated

received signal is deinterleaved in the deinterleaver 288, decoded in the FEC decoder 289, and outputted to the change-over switch 25.

On the other hand, a transmitting signal to  
5 the base station 1 is FEC-encoded in the FEC encoder 291 in the transmitting system 29, interleaved in the interleaver 292, and M-Array encoded in the M-Array encoder 293, as described above. The M-Array-encoded transmitting signal is multiplied by a channelisation  
10 code in the multiplier 294 to be a channel signal, scrambled (spread) by a PN code in the scrambler 297, filtered in the digital filter 296, and converted into an analog signal in the DA converter 297.

This analog signal is orthogonal-modulated in  
15 OQPSK (Offset QPSK) or the like with an RF signal from the FR oscillator 298a in the quadrature modulator 298, filtered in the filter 299, amplified to a necessary transmitting power in the high-power amplifier 300, and transmitted toward the base station 1 through the  
20 hybrid 30a and the transmitting/receiving antenna 30b.

In the terminal structure shown in Fig. 1, the change-over switch 25 switches to (selects) the CDMA transmitting/receiving unit 23 or 24 to be used (namely, a use radio frequency band) under control of the  
25 switching controlling unit 27. The control signal multiplexing/demultiplexing unit 26 has a function of demultiplexing a control signal (selected frequency

notification signal, switching timing instruction signal or the like) contained (multiplexed) in a received signal from the base station 1, and outputting it to the switching controlling unit 27, on the other hand, multiplexing a control signal (reception confirmation signal in response to the selected frequency notification signal, confirmation signal in response to the switching timing instruction signal or the like) to the base station 1 on a transmitting signal to the base station 1.

Namely, the control signal multiplexing/demultiplexing unit 26 fulfills functions (1) to (3) shown below:

(1) function as a selected frequency notification signal receiving unit which receives a selected frequency notification signal for notifying of a radio frequency band selected according to a result of comparison between an own terminal moving speed and the above switching speed threshold value in the base station 1 from the CDMA transmitting/receiving unit 23 or 24 configuring the radio communicating unit;

(2) function as a confirmation signal transmitting unit which transmits a confirmation signal in response to the above selected frequency notification signal to the base station 1; and

(3) function as a switching timing instruction signal receiving unit which receives a switching timing

10043815, 011102



instruction signal in response to the above confirmation signal from the base station 1.

The switching controlling unit (use frequency selection controlling unit; selection controlling unit) 27 controls the change-over switch 25 according to a control signal (selected frequency notification signal, switching timing instruction signal) from the base station 1 demultiplexed from a received signal by the above control signal multiplexing/demultiplexing unit 26 to switch a radio frequency band to be used in the communication with the base station 1.

In Fig. 1, the CDMA transmitting units 16a and 16b, the CDMA receiving unit 16a and 16b, and the CDMA transmitting/receiving units 23 and 24 are provided exclusively for the respective two frequency bands fR1 and fR2. However, it is alternatively possible to provide only parts or circuits (for example, oscillator, synthesizer, filter and the like) required for each of the frequency bands fR1 and fR2, and share parts or circuits other than these, thereby reducing the size of the apparatus.

#### (A3) Description of Operation

Now, description will be made of an operation of the CDMA communication system in the above structure according to this embodiment with reference to Fig. 6.

As shown in Fig. 6, the base station 1 measures the Doppler frequency  $f_D$  on the basis of a received signal by the Doppler frequency measuring unit 20A, and determines a moving speed of the terminal 2 on the basis of the Doppler frequency  $f_D$  by the terminal moving speed estimating unit 20B (step S1). At this time, either the  $f_{R1}$  band or the  $f_{R2}$  band may be used for the communication between the base station 1 and the terminal 2. However, in the initial state, it is desirable to use a lower frequency band (the  $f_{R1}$  band in this embodiment).

In the base station 1, the frequency band switching controlling unit 20D compares the terminal moving speed determined by the terminal moving speed estimating unit 20B with the switching speed threshold value (42.5 km/h in the above case) at predetermined cycles. When the terminal moving speed is not less than the switching speed threshold value, the base station 1 selects the  $f_{R2}$  band (the 2-GHz band, for example) as the use radio frequency band. When the terminal moving speed is less than the switching speed threshold value, the base station 1 selects the  $f_{R1}$  band (the 850-MHz band) as the use radio frequency band (step S2). When the radio frequency bands before and after the selection are the same, selection of the radio frequency band used up to that time is kept.

The frequency band switching controlling unit

20D confirms whether there is an idle channel in the selected frequency band fR1 or fR2. When there is an idle channel, the frequency band switching controlling unit 20D notifies the terminal 2 of the selected  
5 frequency band fR1 or fR2 with a control signal (selected frequency notification signal). When there is no idle channel, a frequency band fR1 or fR2 presently in use is continuously used (steps S3 and S4). In this case, notification of the selected frequency with the  
10 selected frequency notification signal to the terminal 2 may be performed, or not.

When the terminal 2 receives the selected frequency notification signal from the base station 1, the switching controlling unit 27 generates a  
15 reception confirmation signal, and sends it back to the base station 1 (step S5). When receiving the reception confirmation signal, the base station 1 instructs a switching timing of the use radio frequency band to the terminal 2 with a switching timing  
20 instruction signal (step S6). When the terminal 2 receives the switching timing instruction signal, the switching controlling unit 27 generates a confirmation signal, and sends it back to the base station 1 (step S7).

25 After that, the base station 1 and the terminal 2 together wait for a timing defined by the above switching timing instruction signal, and

20140325 01:11:02

simultaneously switch the use radio frequency band thereof to a selected radio frequency band (steps S8 and S9). When radio frequency bands before and after the switching are the same, a radio frequency band used  
5 up to that time is kept, as a result. Synchronization of the switching timing may be established in another manner.

The use radio frequency band to be used between the base station 1 and the terminal 2 is switched to  
10 an optimum frequency band according a terminal moving speed using a communication quality characteristic (frequency dependency) to the Doppler frequency  $f_D$  in the CDMA communication system to improve the worst value of the required  $E_b/N_0$  in the CDMA communication  
15 system in which a relationship between the Doppler frequency  $f_D$  and the transmission quality degradation is non-monotonous, as described above with reference to Fig. 5. This allows improvement of the communication quality and increases the channel  
20 capacity.

#### (B) First Modification

In a radio propagation environment of a mobile communication system such as a CDMA communication system, there generates fluctuation (local median  
25 value fluctuation) in radio wave propagation loss due to shadowing or the like caused by a building or the like (refer to, for example, the above reference 3).

20110528001

Since the local median value fluctuation is log-normal fluctuation, the larger the local median value fluctuation standard deviation ( $\sigma_{stm}$ ) of an interference wave from other cell, the more the other-cell interference ratio ( $f_o$ ) increases and the channel capacity decreases (for example, refer to references 4 and 5 below).

• Reference 4: A. J. Viterbi and A. M. Viterbi, "Other-Cell Interference in Cellular Power-Controlled CDMA", IEEE Trans. On Commun., Vol. COM-42, No. 2/3/4, pp. 1501-1504, (1994).

• Reference 5: A. J. Viterbi, et al., "Soft Handoff Extends CDMA Cell Coverage and Increases Reverse Link Capacity", IEEE J, Selected Areas in Commun., 12(8), pp. 1281-1288.

The higher the radio frequency, the larger the local median value fluctuation standard deviation ( $\sigma_{stm}$ ); the higher the radio frequency, the larger is the other-cell interference ratio  $f_o$ . Since the channel capacity is inversely proportional to  $E_b/N_{o\_thx} (f_s + f_o)$ , as shown by the equation (2), it is possible to more increase the effect of the channel capacity improvement by considering a difference in the other-cell interference ratio  $f_o$  between the radio frequency bands.

For instance, table 3 shows results of determination of the other-cell interference ratios

201105180101 10043815 011102

on forward link in computer simulation where the local median value fluctuation standard deviations in the 850-MHz band and the 2-GHz band are 6dB and 8Db, respectively (refer to the above reference 3).

5 Wherein, conditions of the computation are a distance attenuation parameter = 3.5, shadowing correlation = 0.5, a soft hand-off ratio = 100% and a site ratio = 90%. The self-cell interference ratio is 1 because of forward link.

10 "Distance attenuation parameter" is an index representing how much a power of the radio wave attenuates according to a propagation distance. "Distance attenuation parameter = 3.5" means that a power of the radio wave attenuates of the order of the  
15 negative 3.5-th power. "Shadowing correlation" signifies correlation of interference due to shadowing of the radio wave. The larger the correlation value, the larger is an increase or decrease in power of the radio wave due to the interference of shadowing. The  
20 smaller this value, the smaller is a relationship between an increase or decrease in power of the radio wave and the interference of shadowing. "Soft handoff ratio" represents how many terminals 2 perform soft handoff. For example, when the number of the links  
25 (channels) used for soft handoff is two and one terminal 2 uses these channels, the soft handoff ratio is 100%. "Site ratio" represents a rate of a site in which the

10043815 01102

terminal 2 can actually communicate inside a cell.  
[Table 3]

SIMULATION RESULTS OF  $f_o$  to  $\sigma_{stm}$

| $\sigma_{stm}$ | $f_o$ |
|----------------|-------|
| 6 dB           | 0.88  |
| 8 dB           | 1.26  |

Fig. 7 shows results of calculation of  $E_b/N_{o\_th} \times (f_s + f_o)$  to a terminal moving speed based on the above results. As shown in Fig. 7, when a change in the other-cell interference ratio  $f_o$  is considered, improvement of about 0.7 dB to the worst value (approximately 17.5% when converted into the channel capacity) is expected by switching between the 850-MHz band (the  $f_{R1}$  band) and the 2-GHz band (the  $f_{R2}$  band) at a terminal moving speed 60-80 km/h as the switching speed threshold value, similarly to the above embodiment.

As above, the switching speed threshold value is determined on the basis of a characteristic of "a product of  $E_b/N_{o\_th}$  and an interference power information ( $f_s + f_o$ )" to the radio frequency band and the terminal moving speed, so that a change in the other-cell interference ratio  $f_o$  according to a radio frequency band is considered, which leads to an effect of further improvement of the communication quality and channel capacity.

Meanwhile, structures and operations of the

base station 1 and the terminal 2 are similar to those of the above embodiment, excepting that the switching speed threshold value is determined on the basis of  $E_b/N_o_{th} \times (f_s + f_o)$  to the terminal moving speed as shown in Fig. 7 as above, differently from the above embodiment.

In the above example, the interference power information  $(f_s + f_o)$  is calculated using a local median value fluctuation standard deviation to the radio frequency band. It is alternatively possible to calculate the interference power information  $(f_s + f_o)$  on each radio frequency band adding a change in other wave propagation parameters (for example, refer to the reference 3) such as a distance attenuation parameter to the radio frequency band and the like to increase the effect of the improvement.

#### (C) Second Modification

Local median value fluctuation standard deviation or multi-path interference changes according to operational environments of the system, installation conditions of the base station and the like. For this, values of the self(same)-cell interference ratio  $f_s$  and the other-cell interference ratio  $f_o$  may fluctuate according to operational conditions of the system, installation conditions of the base station or fluctuation in traffic with time.

This embodiment deals with fluctuation in



(fs+fo) as this. When the above equation (2) is modified, the following equation (4) is given.

$$(fs+fo) \doteq K \times \frac{pg}{Eb/No \times N} \quad \dots (4)$$

5

In the equation (4), N is the number of accommodated channels (ch/sector/FR), pg is a processing gain, Eb/No is an actual Eb/No, and K is a coefficient representing an effect of other parameters. Since K and pg are values determined when operational conditions of the system and the like are determined, it is possible to know (fs+fo) when the number of channels K and the actual signal quality Eb/No are determined.

15        The number of channels N is generally a quantity monitored at all times for charging or the like. Eb/No can be readily determined by measuring an error rate or a received power of received data. Accordingly, it is possible to determine an optimum switching speed threshold value under conditions of the actual system operation by beforehand measuring the number of channels and Eb/No in the operational environments of the system or under installation conditions of the base station, calculating (fs+fo), and making a graph  
20  
25        corresponding to Fig. 7 using the value.

Fig. 8 shows a structure of a base station 1A

in this case. A structure of the terminal 2 is similar to that shown in Figs. 1 and 4. Dissimilarly to the base station 1 shown in Fig. 1, the base station 1A according to this modification comprises, in the  
5 controlling system 20, a switching speed threshold value processing unit 20E including a subscriber monitoring unit 20E-1, an Eb/No estimating unit 20E-2, an  $(f_s+f_o)$  calculating unit 20E-3, a switching speed threshold value determining unit 20E-4 and a switching  
10 speed information update controlling unit 20E-5. Incidentally, other structural elements have similar functions shown in Fig. 1 when not specifically mentioned.

The subscriber monitoring unit 20E-1 monitors  
15 the number of channels (the number of subscribers) during communication, which can be realized using a channel monitoring function for charging or the like as described above, for example. The Eb/No estimating unit 20E-2 estimates an actual Eb/No in each radio  
20 frequency band on the basis of error rate information on a received signal and received power information on a received signal in each radio frequency band (the fR1 band, the fR2 band).

The  $(f_s+f_o)$  calculating unit (interference  
25 power ratio information calculating unit) 20E-3 carries out calculation for each radio frequency band using the above equation (4) on the basis of the number

of channels N monitored by the subscriber monitoring unit 20E-1 and an actual measured value of  $E_b/N_0$  obtained by the  $E_b/N_0$  estimating unit 20E-2 to calculate interference power ratio information  
5 (fs+fo) on each radio frequency band. The switching speed threshold value determining unit 20E-4 makes a graph corresponding to Fig. 7 on the basis of a result (fs+fo) of the calculation by the (fs+fo) calculating unit 20E-3 to determine an optimum switching speed  
10 threshold value under actual operational conditions of the system.

The switching speed information update controlling unit 20E-5 applies a start trigger to the (fs+fo) calculating unit 20E-3 at predetermined cycles  
15 to make the (fs+fo) calculating unit 20E-3 calculate (fs+fo) at predetermined cycles.

With such the structure of the base station, the system according to this modification can again calculate an optimum switching speed threshold value,  
20 following fluctuation, even when a value of the self(same)-cell interference ratio fs or the other-cell interference ratio fo fluctuates due to the fluctuation in operational environments of the system, in installation conditions of the base station or in  
25 traffic with time, thereby realizing a suitable frequency band switching in consideration of fluctuation or the like in actual operational

environments of the system, in installation conditions of the base station or in traffic with time. Accordingly, it is possible to largely improve the communication quality and increase the channel capacity.

(D) Third Modification

According to the above embodiment and modifications, the user radio frequency band is switched at one switching speed threshold value (hereinafter simply referred as "threshold value"). For this, when there is deviation in the terminal moving speed, there is a possibility that a phenomenon that one radio frequency band (the fR1 band or the fR2 band) is more frequently selected is generated. According to this modification, there are prepared a plurality of threshold values to alleviate concentration of calls in a specific frequency band.

Fig. 9 shows a structure of a base station 1B according to a third modification. Incidentally, a structure of the terminal 2 according to this modification is similar to that shown in Figs. 1 and 4. Unlike the base station 1 shown in Fig. 1, the base station 1B according to this modification is provided with a frequency band switching controlling unit 20F which accepts a plurality ( $2 \times M - 1$ ) of threshold values, in lieu of the frequency band switching controlling unit 20D.

When  $M=2$ , that is, when two frequency bands  $fR1$  and  $fR2$  ( $fR1 < fR2$ ) are prepared, three threshold values  $V1$ ,  $V2$  and  $V3$  are set as the above threshold value, as shown in Fig. 10.  $V1$  is a terminal moving  
 5 speed at which  $Eb/No$  is at the worst value in the above embodiment and modifications,  $V2$  is a terminal moving speed at which  $Eb/No_{th}$  is at the same value as when the terminal moving speed is  $V1$  in the case where the  $fR2$  band is used, and  $V3$  is a terminal moving speed  
 10 at which  $Eb/No_{th}$  is at the same value as when the terminal moving speed is  $V2$  in the case where the  $fR1$  band is used.

In this case, the frequency band switching controlling unit 20F such controls the change-over  
 15 switches 12, 17 and 19 to use the  $fR1$  band or the  $fR2$  band (using  $fR1$  in preference) when  $VM \leq V1$  where the terminal moving speed is  $VM$ , use the  $fR1$  band when  $V1 < VM \leq 2V$ , use the  $fR2$  band when  $V2 < VM \leq V3$ , and use the  $fR2$  band or the  $fR1$  band (using the  $fR2$  band in preference)  
 20 when  $V3 < VM$ .

Namely, the frequency band switching controlling unit 20F has a function as a determining unit which compares a terminal moving speed detected by the terminal moving speed estimating unit 20A with  
 25 each of  $(2 \times M - 1)$  threshold values, and determines which range of threshold value the terminal moving speed falls in. The radio frequency band is selected on the

basis of a result of determination by the determining unit and priority information defining which radio frequency band should be used for each of plural terminal speed ranges defined by the threshold values.

5           A sequence for switching the use radio frequency band between the base station 1B and the terminal 2 is as shown in Fig. 12. In the base station 1B, the Doppler frequency measuring unit 20A measures the Doppler frequency  $f_D$  on the basis of a received  
10 signal, and the terminal moving speed estimating unit 20B determines a moving speed of the terminal 2 on the basis of the Doppler frequency  $f_D$  (step S1). Incidentally, either the  $f_{R1}$  and  $f_{R2}$  bands may be used in the communication between the base station 1 and  
15 the terminal 2 at this time. However, it is desirable to use the lower frequency band (the  $f_{R1}$  band in this embodiment) in the initial state.

          In the base station 1B, the frequency band switching controlling unit 20F compares the terminal  
20 moving speed determined by the terminal moving speed estimating unit 20B with the threshold value at predetermined cycles. And, the frequency band switching controlling unit 20F selects a use radio frequency band under the above conditions (which range  
25 of the threshold value  $V_1$ ,  $V_2$  or  $V_3$   $V_M$  falls in) shown in Fig. 10. At this time, when there are a plurality of selectable radio frequency bands, the frequency band

10043815 "01102

switching controlling unit 20F selects one having a higher priority (step S2'). When radio frequency bands before and after the selection are the same, selection of a radio frequency band used up to that  
5 time is kept, as well.

The frequency band switching controlling unit 20F confirms whether there is an idle channel in the selected frequency band. When there is an idle channel, the frequency band switching controlling unit 20F  
10 notifies the terminal 2 of the selected frequency band with a control signal (selected frequency notification signal). When there is no idle channel, the frequency band switching controlling unit 20F selects a frequency band having a lower priority, confirms presence of an  
15 idle channel in the similar manner, and notifies the terminal 2 of the selected frequency band with a selected frequency notification signal. When there is no idle channel in any selectable frequency band, a frequency band presently in use is continuously used  
20 (steps S3' and S4). In this case, a frequency band presently in use may be newly notified to the terminal 2 with a selected frequency notification signal, or this notification may be omitted.

After that, the base station 1B and the terminal  
25 2 exchange a reception confirmation signal, a switching timing instruction signal and a confirmation signal in the similar manner shown in Fig. 6, together wait

10043815:011102

for a timing defined by the switching timing instruction signal, and simultaneously switch the use radio frequency band to a selected radio frequency band (steps S5 to S9). When the radio frequency bands before  
 5 and after the switching are the same, a radio frequency band used up to that time is kept as a result, as well. Synchronization of the switching timing may be established in another manner.

With a plurality of threshold values, the use  
 10 radio frequency band is selected and given a priority according to which range of threshold value the terminal moving speed falls in, as above. It is thereby possible to alleviate concentration of calls in a specific frequency band without degrading the worst  
 15 value of  $E_b/N_o\_th$ , similarly to the above embodiment.

When  $M=3$ , that is, when three frequency bands  $fR1$ ,  $fR2$  and  $fR3$  are prepared, five threshold values  $V1-V5$  are set as shown in, for example, shown in Fig. 11. When  $VM \leq V1$ , the  $fR1$  band, the  $fR2$  band or the  
 20  $fR3$  band is used (preferentially in order of the  $fR1$  band and the  $fR2$  band). When  $V1 < VM \leq V2$ , the  $fR1$  band or the  $fR2$  band is used ( $fR1$  being preferentially used). When  $V2 < VM \leq V3$ , the  $fR1$  band is used. When  $V3 < VM \leq V4$ , the  $fR3$  band is used. When  $V4 < VM \leq V5$ , the  $fR3$  band  
 25 or the  $fR2$  band is used (preferentially the  $fR3$  band). When  $V5 < VM$ , the  $fR1$  band, the  $fR2$  band or the  $fR3$  band is used (preferentially in order of the  $fR3$  band and



the fR2 band). This is the same when  $M > 3$ .

Even when a plurality of threshold values are prepared, as this modification, it is possible to determine each threshold value on the basis of

5  $E_b/N_o_{th} \times (f_s + f_o)$  to a terminal moving speed as described with reference to Fig. 7, calculate an other-cell interference ratio  $f_o$  adding a change in another wave propagation parameter (for example, refer to the reference 3) such as a distance attenuation

10 parameter to the radio frequency band or the like, or sequentially update values of the self-cell interference ratio  $f_s$  and the other-cell interference ratio  $f_o$  according to fluctuation in operational environments of the system, in installation conditions

15 of the base station or in traffic with time, similarly to the first and second modifications. In any case, a larger effect of improvement of the communication quality and channel capacity is expected while concentration of calls in a specific frequency band

20 is prevented.

(E) Others

In the above embodiment and modifications, a terminal moving speed itself is detected as information about a terminal moving speed from the Doppler

25 frequency  $f_D$ . However, the Doppler frequency  $f_D$  (fading pitch (cycle) information) itself may be used to set a threshold value for the information. In such

10043815.01.1102

case, the similar functions and effects can be provided, of course.

In the above embodiment and modifications, the base station 1 estimates a state of reception (terminal  
5 moving speed or fading cycle) of a signal on forward link in the terminal 2 on the basis of a received signal (signal on reverse link) from the terminal 2, and switches the use frequency band. Alternatively, the terminal 2 may notify the base station 1 of an actual  
10 state of reception (terminal moving speed or fading cycle) of the terminal 2, whereby the base station 1 determines whether the fading cycle or the terminal moving speed in the received signal on forward link received by the terminal 2 is fast or slow to switch  
15 the frequency in the above manner.

In the above embodiment and modifications, the switching speed threshold value is set to a speed corresponding to an intersection of the required  $E_b/N_0$  characteristics of the radio frequency bands.  
20 Alternatively, it is possible to select, as the use frequency, a frequency belonging to at least the 2-GHz band when the terminal moving speed is below a speed at which the tendency of the required  $E_b/N_0$  characteristic 31 in the 850-MHz band changes in, for  
25 example, Fig. 5, or select, as the use frequency, a frequency belonging to the 850-MHz band when the terminal moving speed is not slower than a speed at

which the tendency of the required  $E_b/N_0$  characteristic  
32 in the 2-GHz band changes.

Note that the present invention is not limited  
to the above examples, but may be modified in various  
5 ways without departing from the scope of the invention.

10043815 011102  
201410 15084007